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TCAD optimization of LGAD sensors for extremely high

- 6 fluence applications
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ABSTRACT: The next generation of high-energy physics experiments at future hadronic colliders 17 will require tracking detectors able to efficiently operate in extreme radiation environments, where 18 expected fluences will exceed $1 \times 10^{17} n_{eq}/cm^2$. This new operating scenario imposes many efforts 19 on the design of effective and radiation-resistant particle detectors. Low-Gain Avalanche Diode 20 (LGAD) represents a remarkable advance because the radiation damage effects can be mitigated by 21 exploiting its charge multiplication mechanism after heavy irradiation. To obtain the desired gain 22 (about 10-20) on the sensor output signal, a careful implementation of the "multiplication" region 23 is needed (i.e. the high-field junction implant). Moreover, a proper design of the peripheral region 24 (namely, the guard-ring structure) is crucial to prevent premature breakdown and large leakage 25 currents at very high fluences, when the bias voltage applied creates an electric field higher than 26 $15 \text{ V/}\mu\text{m}$. In this contribution, the design of LGAD sensors for extreme fluence applications is 27 discussed, addressing the critical technological aspects such as the choice of the active substrate 28 thickness, the gain layer design and the optimization of the sensor periphery. The impact of 29 several design strategies is evaluated with the aid of Technology-CAD (TCAD) simulations based 30 on a recently proposed model for the numerical simulation of radiation damage effects on LGAD 31 devices. 32

33 KEYWORDS: Solid-state silicon detectors, Radiation-hard detectors, Tracking detectors, LGAD,

34 TCAD simulation

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44 1 Introduction

The next generation of High Energy Physics (HEP) experiments, e.g. at the hadronic Future Circular 45 Collider, will require tracking detectors able to efficiently operate in extreme radiation environments, 46 where expected fluences will exceed $1 \times 10^{17} n_{ea}/cm^2$. This new operating scenario imposes many 47 efforts on the design of particle detectors able to be radiation tolerant and to deliver time and position 48 resolutions in the order of a few tens of picoseconds and a few tens of micrometers, respectively. 49 The Low-Gain Avalanche Diode (LGAD) technology represents a remarkable advance because 50 the radiation damage effects can be mitigated by exploiting its charge multiplication mechanism [1] 51 and it offers an intrinsic timing resolution of few tens of picoseconds [2]. To obtain a moderate 52 internal gain (about 10 - 20) and a stable operation, a careful implementation of the gain layer, 53 i.e. the p^+ implant responsible for the signal multiplication, is needed. Moreover, a proper design of 54 the peripheral region (namely, the guard-ring structure) is crucial to prevent premature breakdown 55 and large leakage currents at high fluences. 56

In this contribution, the design and the optimization of LGAD sensors for extreme fluence applications are discussed, addressing the critical technological aspects such as the choice of the active substrate thickness, the design of the gain layer and the sensor periphery. The impact of several design strategies is evaluated with the aid of Technology-CAD (TCAD) simulations based on a recently proposed model for the numerical simulation of radiation damage effects on LGAD devices [3].

2 TCAD model for the numerical simulation of LGAD sensors

TCAD tools can be proficiently used to evaluate in advance the effects of layout and technological parameters on the device performance before its production. In this work, the analyses on existing structures and the new developments in the LGAD design have been done following detailed device-level simulations by means of the state-of-the-art Synopsys Sentaurus TCAD platform [4].

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68 2.1 Layout and doping profile

The simulated device has been initially designed by neglecting the edges of the multiplication implant to focus the study on avalanche effect due to the high electric field generated by the gain

⁷¹ layer. Moreover, the high sensitivity of the gain layer to its small technological variation has required

⁷² a stringent mesh refinement and thus the need for a "quasi-1D approach", as depicted in figure 1.

The highly doped n-type strip and the moderately doped p-type gain layer have been modeled by means of a Gaussian analytical profile. Moreover, to take into account the acceptor-removal mechanism that occurs after the irradiation [5], the doping profile of the multiplication layer has

⁷⁶ been properly reduced for increasing values of fluence, by using the following analytical law:

$$N_A^{peak}(\Phi) = N_A^{peak}(0) \cdot e^{-c_A \cdot \Phi} .$$
(2.1)

⁷⁷ According to that, the peak dose of the gain layer, N_A^{peak} , is recomputed as a function of the initial ⁷⁸ peak acceptor density, $N_A^{peak}(0)$, and the exponential dependence with the fluence, Φ , and the ⁷⁹ acceptor removal factor, c_A , which is calculated from the "Torino parameterization" [6].

80 2.2 Simulation outcomes and model validation

The steady-state behavior of the device has been simulated under different bias voltage and fluence 81 conditions. The current-voltage curves of the irradiated devices have been obtained at 253 K and 82 then scaled to 300 K to consider the temperature dependence of the current generated in the silicon 83 bulk [7] and to have a direct comparison with experimental data [8]. Special attention has been 84 devoted to the choice of the avalanche model, by investigating among the embedded available 85 ones (e.g. Van Overstraeten-De Man [9], Okuto-Crowell [10] and University of Bologna [11]) as 86 well as by extending the TCAD "portfolio" by adding different ones (e.g. Massey model [12]) 87 through the code library customization [13]. Figure 2 top left shows significant differences between 88 the models in terms of steady-state behavior. These are due to the different value of the impact 89 ionization coefficients used in each model, and according to that, the Massey model presents the 90 best agreement with the experimental data. In order to have a predictive insight into the electrical 91 behavior and the charge collected by the LGADs up to the highest particle fluences expected in 92 the future HEP experiments, the well validated "New University of Perugia" radiation damage 93 model has been implemented within the TCAD simulation environment [14][15]. By coupling this 94 numerical model, which allows to consider the comprehensive bulk and surface damage effects 95 induced by radiation on silicon sensors, with the analytical law that describes the mechanism of 96 acceptor removal in the gain layer, it has been possible to reproduce experimental data with high 97 accuracy (see figure 2), demonstrating the reliability of the implemented simulation framework. 98

3 LGAD design and optimization

The good agreement obtained between simulation results and measurement data has allowed us to apply the newly developed model not only for the analysis of the device behavior, but also for the design and optimization of the future productions of LGAD sensors, considering their possible use in the next generation of collider experiments. To this purpose, in section 3.1 we compare different technology solutions in terms of substrate thickness and gain layer implant, and in section 3.2 the performances of different peripheral region designs are analyzed.



Figure 1. On the left, the layout of the LGAD device implemented within the TCAD environment. On the right, the simulated doping profiles. The n^{++} -strip and the p^+ -gain implant are modeled by means of a Gaussian analytical profile. The gain layer doping profile is shaped according to the analytical law in eq. (2.1) to take into account the acceptor removal mechanism after the irradiation.



Figure 2. Comparison between simulated curves (Sim.) and experimental data (Meas.) carried out on a 55 μ m-thick, 1 × 1 mm² LGAD device at 300 K [8]. Top left: current-voltage curves before irradiation for different avalanche models. Top right: current-voltage curves before and after irradiation. Bottom left: gain-voltage curves before irradiation. Bottom right: gain-voltage curves at the fluence of $1.5 \times 10^{15} n_{eq}/cm^2$.

106 3.1 The gain layer implant

Since the technological parameters of the gain layer implant strongly influence the multiplication 107 capability of an LGAD device, different combinations of peak doping concentration, implantation 108 depth and width have been investigated. In particular, three gain layer profiles have been designed 109 and called "Shallow", "Standard" and "Deep" (see figure 3 left). For simplicity, these doping 110 profiles are labeled with the letters A, B and C, and they are represented in red, green and blue, 111 respectively. Figure 3 right reports the comparison between the simulated current-voltage curves 112 for each considered gain layer doping profile for not irradiated devices. Differences in terms of 113 depletion and breakdown voltage (V_{BD}) are clearly visible. For example, the "Deep" profile is 114 characterized by a higher gain layer depletion voltage and a lower breakdown with respect to the 115 other profiles, thus implying a smaller operating range of voltage. Moreover, a careful tuning of the 116 peak dose of the gain layer implant has been done and a safety value of the breakdown voltage has 117 been set at about 200 V to prevent the early breakdown of the device. 118

Once the peak dose of each gain layer profile has been fixed for not irradiated devices, the 119 analysis focuses on the steady-state behavior and the gain evolution of the devices simulated at high 120 irradiation fluences. Figure 4 left shows that the devices with different gain layer profiles behave 121 like a PIN diode at the highest value of simulated fluence, i.e. $5 \times 10^{16} n_{eq}/cm^2$. The typical "knee" 122 trend and the relatively high slope of the current-voltage curves before the breakdown are almost 123 absent at that fluence. This implies that the LGAD device lost its multiplication power, despite the 124 presence of active dopant, as confirmed by the unitary value of the gain-voltage curves in figure 4 125 right (see the cross markers): the remaining $N_A^{peak}(\Phi)$ does not allow for impact ionization to occur. 126 Nevertheless, the simulations reveal that the signal multiplication capability is still preserved at a 127 fluence of $1 \times 10^{16} n_{ea}/cm^2$. The gain-voltage curve of the "Shallow" profile (red square markers) 128 shows that the value of the gain is roughly equal to ten when the sensor is biased at around 350 V 129 and it is even equal to forty when the sensor is biased at the breakdown voltage ($V_{BD} \approx 380$ V). 130 This is a hint of strong resilience of the "Shallow" profile to the radiation damage thanks to its 131 higher residual acceptor density with respect to the other profiles. 132

Finally, sensors characterized by different substrate thicknesses have been simulated with the aim of investigating their electrical behavior once a specific gain layer doping profile design has been chosen (e.g. "Shallow"). The current-voltage curves of $15 \,\mu$ m, $30 \,\mu$ m and $45 \,\mu$ m-thick LGAD sensors are represented in figure 5 on the left, center and right, respectively. As mentioned before, by properly tuning the peak dose of the gain layer profile, it is possible to fix the breakdown voltage at a chosen value.

3.2 The sensor periphery

In figure 6, an example of a sensor periphery structure implemented within the TCAD environment is reported. The structure is composed by a collector ring, i.e. the bias-ring, a floating guard-ring (GR), and a scribe line. In particular, the floating GR has been specifically devised with both an n-deep and a p-stop implant. The latter helps to cut off the possible build-up of the so-called "inversion layer", i.e. a surface leakage current path between the collector ring and the n-deep implant itself, thus avoiding a direct connection between them [16].



Figure 3. Gain layer sensitivity analysis. Impact of three different doping profiles (called "Shallow", "Standard" and "Deep" - on the left) on the simulated steady-state behavior of a 20 μ m-thick LGAD device before irradiation at 300 K (see the current-voltage curves on the right). A variation of a few percentages of the gain layer peak dose is sufficient to produce a significant change of the breakdown voltage (V_{BD}).



Figure 4. On the left, comparison between the steady-state behavior of the three gain layer profiles, and on the right, comparison between the corresponding gain-voltage curves simulated at the fluence of $1 \times 10^{16} \, n_{eq}/cm^2$ (square markers) and $5 \times 10^{16} \, n_{eq}/cm^2$ (cross markers).



Figure 5. Steady-state behavior of a $15 \,\mu\text{m}$ (left), $30 \,\mu\text{m}$ (center), and $45 \,\mu\text{m}$ (right) thick LGAD sensor before irradiation at 300 K and tuning of the breakdown voltage by varying the peak dose of the gain layer ("Shallow" profile).

To allow the sensor to effectively operate in a harsh radiation environment, the optimization of 146 the GR structure is a crucial task, specially when small substrate thicknesses are used. In figure 6 147 right, the V_{BD} trend as a function of a 15 µm, 25 µm, 35 µm and 45 µm-thick active substrate layer 148 (A_L) before and after irradiation is shown. Not only the V_{BD} decreases by reducing A_L , as expected, 149 but also the difference between the V_{BD} before and after irradiation decreases. The electrostatic 150 potential and the electric field are represented in figure 7 left and right, respectively. They are 151 both evaluated at a bias equivalent to V_{BD} and at a fluence of $1 \times 10^{15} \, n_{eq}/cm^2$ as a function of 152 the position along the surface region at a depth of 100 nm underneath the silicon-oxide interface. 153 The sharp drop of the electrostatic potential in proximity of the implants implies high-electric-field 154 peaks ($E_f \approx 40 \,\mathrm{V/\mu m}$) which can trigger a premature breakdown. To ensure the stability and 155 uniformity of the electric field distribution, several GR design strategies have been simulated (see 156 table 1). As an example, as depicted in figure 8 left, a floating GR placed at a given distance from 157 the collector ring allows to fix the voltage along the ring itself at an intermediate value, causing 158 the redistribution of the electrostatic potential to a larger area and thus a significant reduction of 159 the electric field, resulting in a higher V_{BD} . Moreover, by exploiting the concurrent action of bulk 160 and surface damage to mitigate the creation of the inversion layer below the silicon-oxide interface 161 [17], it is possible to design the periphery structure without p-stop isolation. As shown in figure 8 162 right, the V_{BD} is higher when the floating GRs do not have any p-stop implant, which might float 163 to a potential value quite close to the bias, thus implying non-sustainable large voltage drops. 164

Table 1 . Name and description of the simulated GR structure

GR design	Description
full	1 or 3 floating GRs, with n-deep and p-stop
n-deep ONLY (X)	1 or 3 floating GRs, without p-stop implant
	(n-deep at $X \mu m$ distance from the bias-ring)
p-stop ONLY (cont.)	1 or 3 floating GRs, without n-deep implant
	(with an electric contact over the p-stop)



Figure 6. On the left: an example of a sensor periphery layout implemented within the TCAD environment. It consists of a collector ring, a floating guard-ring (i.e. the n-deep implant coupled with the p-stop one), and a scribe line. On the right: trend of the breakdown voltage, V_{DB} , as a function of the active substrate thickness, A_L , before and after irradiation ($\Phi = 1 \times 10^{15} \text{ n}_{eq}/\text{cm}^2$).



Figure 7. Simulated curves of electrostatic potential (on the left) and electric field (on the right), evaluated after the irradiation ($\Phi = 1 \times 10^{15} n_{eq}/cm^2$) at the breakdown voltage, expressed as a function of the position along the surface region of the sensor periphery, 100 nm underneath the silicon-oxide interface, for different values of the active substrate thickness (15 µm, 25 µm, 35 µm and 45 µm).



Figure 8. V_{BD} values obtained before and after irradiation ($\Phi = 1 \times 10^{15} n_{eq}/cm^2$) by simulating a 45 µmthick, 500 µm-wide sensor periphery structure characterized by one floating GR (graph on the left) and three floating GRs (graph on the right), for different combinations of GR design strategies (see table 1).

165 4 Conclusions

In this work, the design and the optimization of LGAD sensors for extreme fluence applications, 166 e.g. future collider experiments, have been presented. The impact of several design strategies has 167 been evaluated with the aid of TCAD simulations based on a recently proposed model for the 168 numerical simulation of radiation damage effects on LGAD sensors. Guidelines for a new production 169 have been identified: i) a gain layer implant characterized by a highly-peaked and narrow high-field 170 junction close to the n^{++} -strip electrode exhibits high residual acceptor density after a fluence of 171 $1 \times 10^{16} n_{eq}/cm^2$, i.e. high radiation tolerance; *ii*) different substrate thicknesses will be used, 172 e.g. $15 \,\mu\text{m}$, $20 \,\mu\text{m}$, $35 \,\mu\text{m}$ and $45 \,\mu\text{m}$, for which the impact of the technological parameters of the 173 gain layer on the breakdown voltage has been evaluated in advance; *iii*) a sensor periphery with 174 floating GR seems to be more effective without any p-stop implant, benefiting from the concurrent 175 action of bulk and surface damage to mitigate the build-up of the inversion layer; iv) by properly 176 positioning the floating GR(s) between the collector ring and the scribe line, it is possible to increase 177 the V_{BD} , thus avoiding the occurrence of early breakdown. 178

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