

# Design optimization of the UFSD inter-pad region

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## Abstract

This paper reports on a measurement campaign to characterize the inter-pad region of Ultra-Fast Silicon Detectors (UFSDs) manufactured by Fondazione Bruno Kessler. The devices under test are either pixel or strip arrays, featuring a large number of different inter-pad layouts; both pre-irradiation and irradiated sensors have been measured. The aim of the study is to link the design parameters of the inter-pad region to the operation of the sensors, providing insights into the design of UFSD arrays with narrow inter-pad gaps. We concluded that, in the UFSD design, the doping level and the area of the  $p$ -stop should be kept low, in order to avoid the early breakdown of the device and the micro-discharges effect; UFSDs with such characteristics proved also rather insensitive to floating pads and irradiation. Thanks to these findings, it was possible to design a UFSD array that yields the expected performance with an inter-pad width as small as 25  $\mu\text{m}$ , significantly improving its fill factor with respect to standard designs. Two innovative experimental techniques are presented in this work: the first one is based on a TCT setup, the second makes use of an ultra-low light CCD camera.

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## 1 **1. Introduction**

2 Ultra-Fast Silicon Detectors (UFSDs) [1] are well-established timing detec-  
3 tors based on the Low-Gain Avalanche Diode (LGAD) technology [2]. UFSDs  
4 have been chosen to instrument the timing layers of the ATLAS and CMS de-  
5 tectors [3, 4], and are envisioned for future 4D-trackers [5, 6], namely detectors  
6 able to concurrently measure the position and time of passage of ionizing parti-  
7 cles with  $\sim 10 \mu\text{m}$  and  $\sim 10 \text{ps}$  resolution, respectively. In UFSD pixel arrays,  
8 signal multiplication happens only inside the pixel, while the area in between  
9 pixels, called inter-pad or no-gain region, does not provide signal multiplica-  
10 tion. An important line of research in the evolution of the UFSD design aims  
11 at maximizing the fill factor, i.e., the ratio between a sensor active and total ar-  
12 eas [7, 8, 9, 10]. This paper reports an in-depth study of inter-pad designs using  
13 sensors of the UFSD3 and UFSD3.1 productions manufactured by Fondazione  
14 Bruno Kessler (FBK, Italy).

## 15 **2. The FBK UFSD3 and UFSD3.1 productions**

### 16 *2.1. UFSD3*

17 The third UFSD production by FBK consists of twenty 6-inch wafers (ac-  
18 tive thickness  $55 \mu\text{m}$ ) with five splits of gain dose and a variety of different  
19 geometries (strip arrays, 2x2 and 5x5 pixel matrices, single pads). The wafers  
20 were produced using the stepper technology instead of the mask-aligner tech-  
21 nique previously used. This choice was made considering the better precision  
22 and higher yield offered by the stepper process.

23 In UFSD3, four different inter-pad designs have been implemented: *Aggres-*  
24 *sive*, *Intermediate*, *Safe*, and *Super-Safe*, with a nominal distance between gain  
25 layers of  $\sim 10 \mu\text{m}$ ,  $\sim 20 \mu\text{m}$ ,  $\sim 30 \mu\text{m}$ , and  $\sim 40 \mu\text{m}$ , respectively; as a  
26 reference, the inter-pad width of the previous production, UFSD2, is  $70 \mu\text{m}$ .  
27 Table 1 lists these variations.

Table 1: Inter-pad design in the UFSD3 production

Name	distance [ $\mu\text{m}$ ]
Aggressive	$\sim 10 \mu\text{m}$
Intermediate	$\sim 20 \mu\text{m}$
Safe	$\sim 30 \mu\text{m}$
Super-Safe	$\sim 40 \mu\text{m}$
Reference - UFSD2	$\sim 70 \mu\text{m}$

28 *2.2. UFSD3.1*

29 The UFSD3.1 production was developed to study the interplay between the  
 30  $p$ -stop <sup>1</sup> doping levels and different layouts of the inter-pad area. It consists  
 31 of seven 6-inch epitaxial wafers with an active thickness of 55  $\mu\text{m}$ . All wafers  
 32 have the same gain layer dose while they differ for the  $p$ -stop doping (Table 2),  
 33 with doping 1 (in arbitrary unit) being the value used in UFSD2 and UFSD3.  
 34 Since the goal is the comparative study of the inter-pad design, it was decided  
 35 to have only one sensor layout consisting of a  $2 \times 2$  pad matrix, with a pad  
 36 size of  $1.3 \times 1.3 \text{ mm}^2$ . UFSD3.1, as UFSD3, has been produced using the stepper  
 37 lithographic technology.

38 The variations of the UFSD3.1 layout focused on two aspects:

- 39 1. The layout of the inter-pad region.
- 40 2. The shape of the regions where  $p$ -stops cross each other.

41 Figure 1 graphically describes the five designs under study:

- 42 1. Grid:  $p$ -stops form a grid around the pads
- 43 2. Full: a  $p$ -stop grid with full disks of  $p$ -doping where  $p$ -stops cross each  
 44 other.

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<sup>1</sup>A  $p$ -stop is a  $p^+$ -doped structure implanted in the inter-pad region to isolate two adjacent pads.

- 45 3. Empty: a  $p$ -stop grid with empty disks of  $p$ -doping where  $p$ -stops cross  
 46 each other.
- 47 4. 2  $p$ -stops: each pixel has its own  $p$ -stop.
- 48 5. Grid guard-ring: the guard-ring forms a grid, isolating each pixel. Each  
 49 pixel has its own  $p$ -stop .

50 These five designs have been implemented with different inter-pad widths  
 51 and sizes of the regions  $A$  and  $B$ , for a total of 11 different inter-pad structures,  
 52 listed in Table 3.

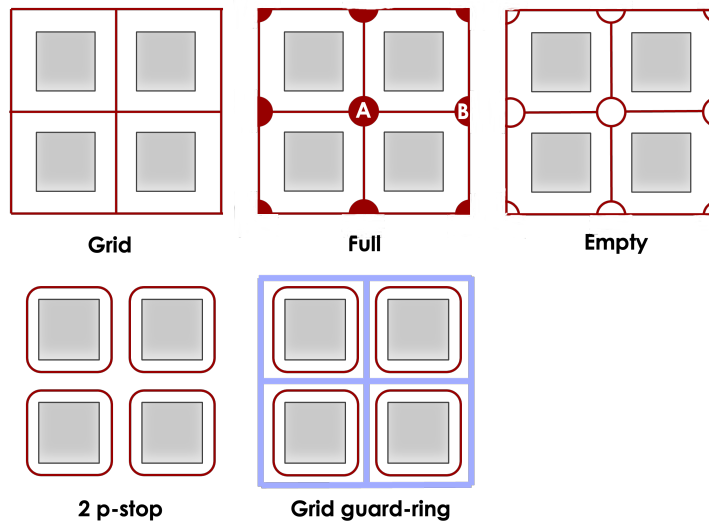


Figure 1: Sketch (not to scale) of the structures present in the inter-pad region of the UFSD3.1 design. The  $p$ -stops are shown in maroon, and the guard-ring is in blue. The areas where two  $p$ -stops cross each other ( $A$  and  $B$  in the picture) are particularly critical as they might lead to premature breakdown.

53 In the *grid* design, the  $p$ -stops cross each other at a 90 degrees angle, whereas  
 54 in the *full* and *empty* designs, different radii of curvature are used, with large  
 55 curvature radii resulting in large areas and smooth corners and small radii yield-  
 56 ing small areas and sharper corners.

Table 2: Wafers of the UFSD3.1 production

Wafer number	$p$ -stop doping [a.u.]
W12	0.02
W13	0.05
W14	0.1
W16	0.15
W17	0.2
W18	1

Table 3: Inter-pad design of the UFSD3.1 arrays.

Type	Nominal inter-pad width [ $\mu\text{m}$ ]	<i>Region A/B</i> design	<i>Region A</i> area [a.u.]	<i>Region B</i> area [a.u.]
1	16	Grid	A	2B
2	20.5	Full	100A	10B
3	20.5	Full	10A	B
4	23.5	Grid	2A	2B
5	25	Full	20A	2B
6	27.5	Full	10A	B
7	27.5	Full	20A	2B
8	27.5	Grid	2A	2B
9	38	2 $p$ -stops		
10	49	Grid guard-ring		
11	25	Empty	100A	10B

### 57 **3. Experimental Techniques**

58 Two experimental techniques have been employed in this work: the Transient  
 59 Current Technique (TCT), and CCD thermal camera imaging.

60 *3.1. The TCT Setup*

61 The TCT technique exploits the motion of non-equilibrium  $e-h$  pairs created  
62 by a laser entering the Device Under Test (DUT). The free charge carriers drift  
63 towards the electrodes, inducing a current signal which is a function of time.  
64 The analysis of the evolution in time of such current signal is the basis of the  
65 TCT technique, providing a wide number of detailed information about the  
66 DUT. For this analysis, the TCT set-up produced by Particulars [11, 12] was  
67 used.

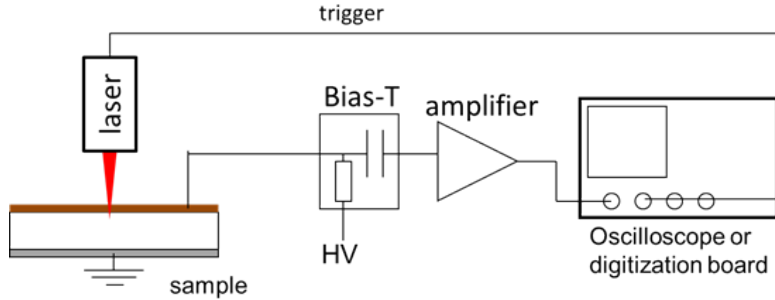


Figure 2: Principle of operation of the Transient Current Technique setup.

68 The basic scheme of a TCT system is shown in fig. 2: the laser lights the  
69 sensor, producing a signal that is amplified by an external trans-impedance  
70 amplifier (a 40 dB Cividec amplifier with 2 GHz bandwidth) and then fed to  
71 a fast oscilloscope (a Teledyne-Lecroy HDO9404 oscilloscope with a 20 GS/s  
72 sampling rate ), where it is recorded for the offline analysis.

73 An important feature of the Particulars system is the possibility of mount-  
74 ing the DUT on a translator x-y stage (Standa 8MTF-102LS05/8MT175-100),  
75 which can be moved with sub-micron precision over a range of tens of centime-  
76 ters; in this way, the laser shot position can be accurately chosen, and the whole  
77 surface of the DUT can be mapped.

78 The laser usually employed in the UFSD characterization is infrared with a  
79 wavelength  $\lambda = 1060$  nm: its absorption depth in silicon is about 1 mm [13],  
80 so it manages to fully cross the DUT, uniformly creating charges along its path

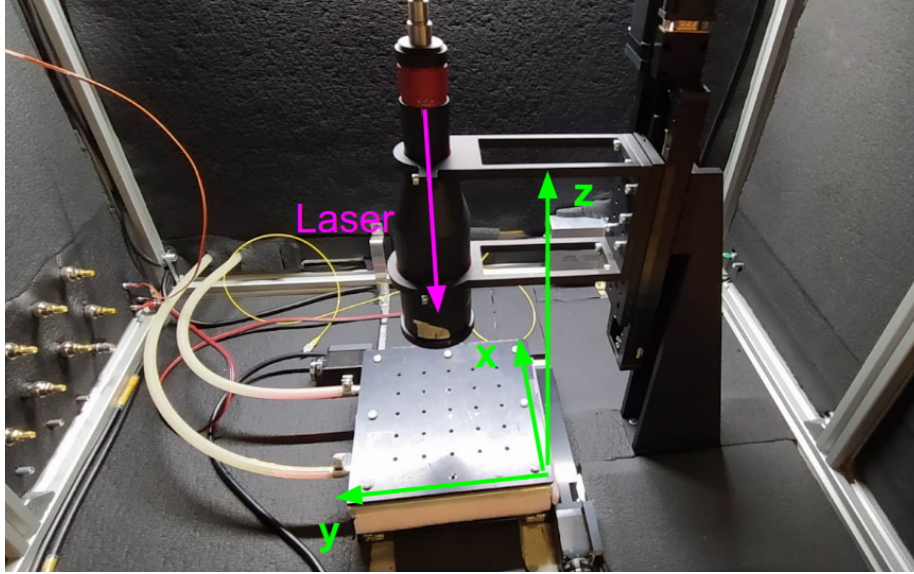


Figure 3: TCT setup in the Torino Laboratory for Innovative Silicon Detectors.

81 and, thus, well simulating the passage of a MIP.

82 The lasers used in this work, produced by Particulars, are single-mode pulsed  
83 lasers with a core diameter of  $\sim 6 \mu\text{m}$ . The laser intensity can be varied  
84 depending on the measurement: it can be set to a level corresponding to a  
85 signal lower than that generated by a MIP, up to many hundreds of MIPs.

86 The lasers used in this work have frequencies in the range 50 Hz - 1 MHz  
87 and their pulse durations range from  $\sim 50$  ps to 4 ns, with symmetrical pulses.

### 88 3.2. The CCD thermal camera

89 A CCD thermal camera is able to perform ultra-low light imaging and it  
90 exploits photon emission to identify the location where the density of current is  
91 high.

92 For this work, the EM-CCD camera Hamamatsu ORCA2 C11090-22B was  
93 used [14]. It consists of a  $1024 \times 1024$  pixel matrix providing 1M pixel resolution  
94 and high quantum efficiency from near-IR to UV, and it is particularly suitable  
95 for applications requiring long exposure times and low noise.

96 In this work, the camera is employed to take pictures of the UFSD inter-pad  
97 region. For such purpose, the camera is mounted on top of the microscope of a  
98 probe station using a specific ocular, as shown in 4; this allows examining the  
99 inter-pad region in detail.

100 An external module driver and dedicated software control the camera. When  
101 the camera is turned on and positioned on the ocular, it can display the field of  
102 view and take pictures. Different exposure times can be chosen, and the camera  
103 can be run both in *normal* mode (a standard CCD camera) and *gain* mode, in  
104 which the light input is multiplied by a specific gain factor that can be chosen  
105 by the user. The latter mode is useful when seeking to detect very faint light  
106 sources. The acquired picture can be either displayed in *black & white* or in  
107 2-/3- colors mode. The light wavelength range in which the camera operates  
108 changes dynamically by default, but the user can also set a fixed range.

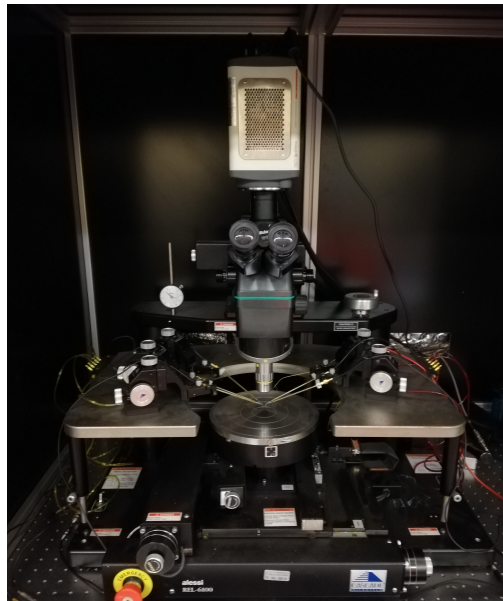


Figure 4: The ORCA2 camera mounted on the probe station of the Torino Laboratory of Innovative Silicon Sensors.

109 During the measurements presented in this work, performed at room tem-  
110 perature, the DUT is placed on the probe station metal chuck, which provides



111 the bias voltage on the backside, while the guard-ring and the pads on the front  
 112 side are contacted with needles and grounded. Then the camera is positioned  
 113 on the ocular, and the microscope is manually focused, with a magnification  
 114 such that all the regions of interest can be viewed.

#### 115 4. Inter-pad widths of the UFSD3 and UFSD3.1: nominal vs mea- 116 sured

117 The inter-pad width of the sensors presented in this work has been measured  
 118 in the laboratory, using the TCT setup. The sensors tested have a small region  
 119 without metal traces (optical window) from one pad to the neighboring one,  
 120 specifically designed for this measurement.

121 The measurement is made by performing a TCT scan between two adjacent  
 122 pads and acquiring their collected charges as a function of the laser position.  
 123 For each sensor to be evaluated, the scan is performed a hundred times. The  
 124 uncertainty on the measured width is about  $2 \mu\text{m}$ .

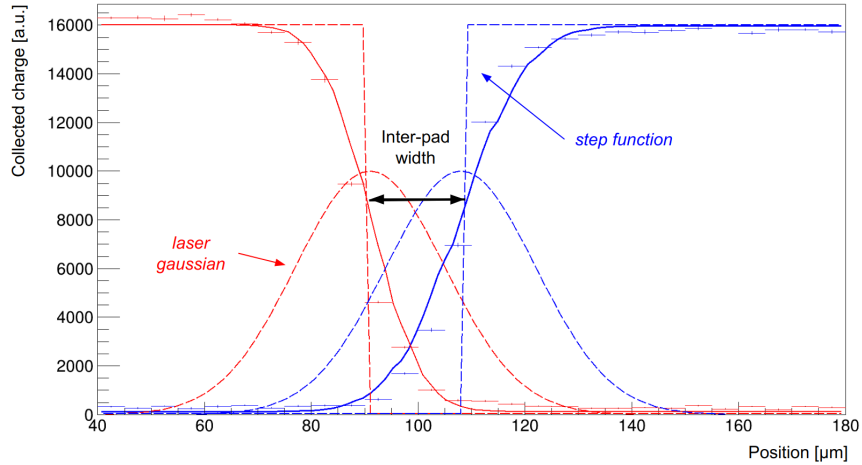


Figure 5: Collected charge as a function of position of two neighboring pads. The step function and the gaussian profile result in a sigmoidal function which is used for the inter-pad width measurement.

125 The expected charge profile from the acquisition consists of two sigmoidal

126 functions (figure 5), one for each read-out pad. The sigmoid is obtained by the  
 127 convolution of a step function (describing the transition between gain and no  
 128 gain regions) with a gaussian function, which accounts for the laser beam spot  
 129 size. The point at which the sigmoid reaches 50% of its height corresponds to  
 130 the intercept with the step function used in the convolution: hence, the width  
 131 of the no-gain region is given by the distance between the 50% points of the  
 132 charge profiles of the two pads.

133 Table 4 presents the measured and nominal inter-pad widths of the sensors  
 134 tested in this work. A detailed description of the inter-pad width measurement  
 135 can be found in [15, 16].

Table 4: Inter-pad widths measured with the TCT setup. The uncertainty on the measured widths is  $\pm 2 \mu\text{m}$ .

Production	design	measured inter-pad [ $\mu\text{m}$ ]	nominal inter-pad [ $\mu\text{m}$ ]	Thermal load
UFSD2		67	70	Low
UFSD3	<i>Aggressive</i>	16.5	11	Low
UFSD3	<i>Medium</i>	16.5	20.5	Low
UFSD3	<i>Medium</i>	31	20.5	High
UFSD3	<i>Safe</i>	30.5	31	Low
UFSD3	<i>Super-Safe</i>	38	41	Low
UFSD3.1	Type 1	32	16	High
UFSD3.1	Type 10	62	49	High
UFSD3.1	Type 11	36	20.5	High

136 As reported in [17], the measured inter-pad distance is always larger than  
 137 the nominal one due to the bending of the electric field lines caused by the  
 138 presence of the JTE structures around the pixel periphery. It is also interesting  
 139 to note that sensors produced with a higher thermal load have a wider inter-pad  
 140 distance.

141 **5. The inter-pad design of the FBK UFSD3 production**

142 The UFSD3 production aimed at exploring the stability of designs with  
 143 nominal inter-pad widths in the range 11 - 49  $\mu\text{m}$ , considerably reducing the  
 144 values reached in the UFSD2 production ( $\sim 70 \mu\text{m}$ ).

145 The first step of the UFSD3 characterization was to measure the  $I(V)$  char-  
 146 acteristic of UFSD arrays with different designs, reported in Figure 6. *Super-safe*  
 147 sensors, namely sensors with the largest inter-pad width, have the typical  $I(V)$   
 148 curve of a sensor with gain: it follows an exponential trend till breakdown, in  
 149 this case above 300 V.

150 Since all measured sensors have the same gain layer doping, the breakdown  
 151 voltage ( $V_{BD-Gain}$ ) due to gain is common to all. However, the *Safe*, *Medium*  
 152 and *Aggressive* designs suffer from premature (or early) breakdown, with an  
 153 abrupt, not exponential,  $I(V)$  characteristic. In particular, Figure 6 shows that  
 154 the narrower the inter-pad width, the earlier  $V_{BD}$ . The abrupt  $I(V)$  curves  
 155 indicate that the breakdown is not caused by an avalanche in the gain region.

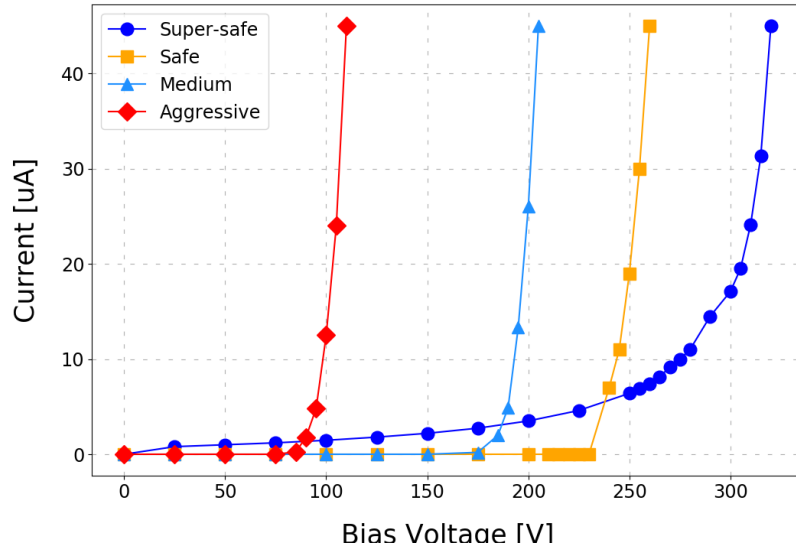


Figure 6:  $I(V)$  curves of UFSD3 sensors with different inter-pad widths

156 Since the only difference among tested sensors is the inter-pad design, the  
157 early breakdown likely originates there. The steps taken to understand the  
158 origin of the early breakdown are explained in the following part of this section.  
159 The samples used for the study are:

- 160 • A strip sensor with *Super-safe* inter-pad design (600  $\mu\text{m}$  pitch, 1 cm  
161 length)
- 162 • A  $2\times 2$  pad matrix with *Safe* inter-pad design ( $1\times 3\text{ mm}^2$  pads)
- 163 • A  $2\times 2$  pad matrix with *Medium* inter-pad design ( $1\times 3\text{ mm}^2$  pads)
- 164 • A strip sensor with *Medium* inter-pad design (300  $\mu\text{m}$  pitch, 1 cm length)
- 165 • A  $2\times 2$  pad matrix with *Aggressive* inter-pad design ( $1\times 3\text{ mm}^2$  pads)

166 The working hypothesis has been that the early breakdown happens some-  
167 where in the inter-gap region, and in this spot, a high electric field should be  
168 present. This high electric field generates local multiplication of the charge  
169 carriers, yielding localized gain and photon emission.

170 All DUTs are covered with a metal layer over the gain region, with the  
171 exception of the optical windows used for the inter-pad measurement. Therefore,  
172 only the no-gain and optical window areas could be scanned with the TCT setup.  
173 The measurements have been performed at room temperature, with the sensor  
174 bonded to a custom read-out board. Each read-out channel is connected to a  
175 40-dB external broadband amplifier, whose output is then connected to a fast  
176 oscilloscope.

177 The DUTs are firstly scanned at a voltage well below breakdown to record  
178 the collected charge and produce an x-y map that clearly defines the DUT and,  
179 in particular, the inter-pad region. As an example, Figure 7 on the left shows the  
180 layout of the sensor, while on the right the TCT collected charge (sensor with  
181 *Safe* design). The z-axis (color-coded) reports the sum of the charges collected  
182 by all four pads. The gain regions covered by the metal have a collected charge  
183 close to zero (in blue).

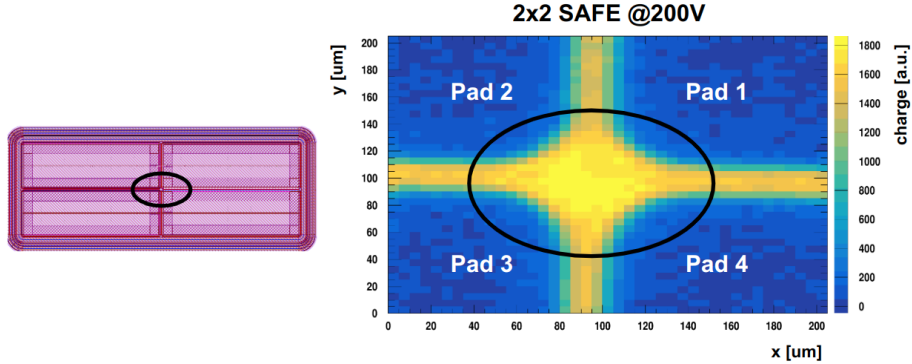


Figure 7: Left: sensor layout, right: 2D-map of the charge collected in a sensor with *Safe* design.

184 The three plots of Figure 8 show, for Pad 1, the evolution of the collected  
 185 charge as a function of the bias voltage. The charge is constant across the  
 186 inter-pad region at 200 V; it increases significantly at 250 V, and at 260 V, the  
 187 breakdown begins. Figure 9 shows the x-projection of these 2D-maps at  $y=100$   
 188  $\mu\text{m}$  where the current increase with bias at  $x = 120 \mu\text{m}$  signals the onset of  
 189 the breakdown.

190 Figure 10 shows a similar result for the strip sensor with *Medium* design:  
 191 the charge collected around the strip corner ( $y = 150 \mu\text{m}$ ) is constant at 200 V,  
 192 then increases approaching breakdown.

193 The *Super-safe* sensor, instead, does not show any increase in collected  
 194 charge in the inter-pad region when brought into breakdown, as shown in 11:  
 195 this is the only design able to reach the expected breakdown voltage  $V_{BD-Gain}$ .

196 Hence, sensors suffering from premature breakdown show signs of charge  
 197 multiplication occurring in the inter-pad region, whereas the *Super-safe* devices  
 198 have a constant collected charge in the inter-pad up to breakdown. The charge  
 199 multiplication in the inter-pad region is caused by the onset of a strong electric  
 200 field, underlying a weakness in the design. The effect of such a field appears  
 201 suddenly, as proven by the trend of the  $I(V)$  curves in those devices.

202 The high electric field occurring in the inter-pad region can be explained as  
 203 due to the  $pn$  junction between the  $p$ -stop and the *inversion layer* underneath

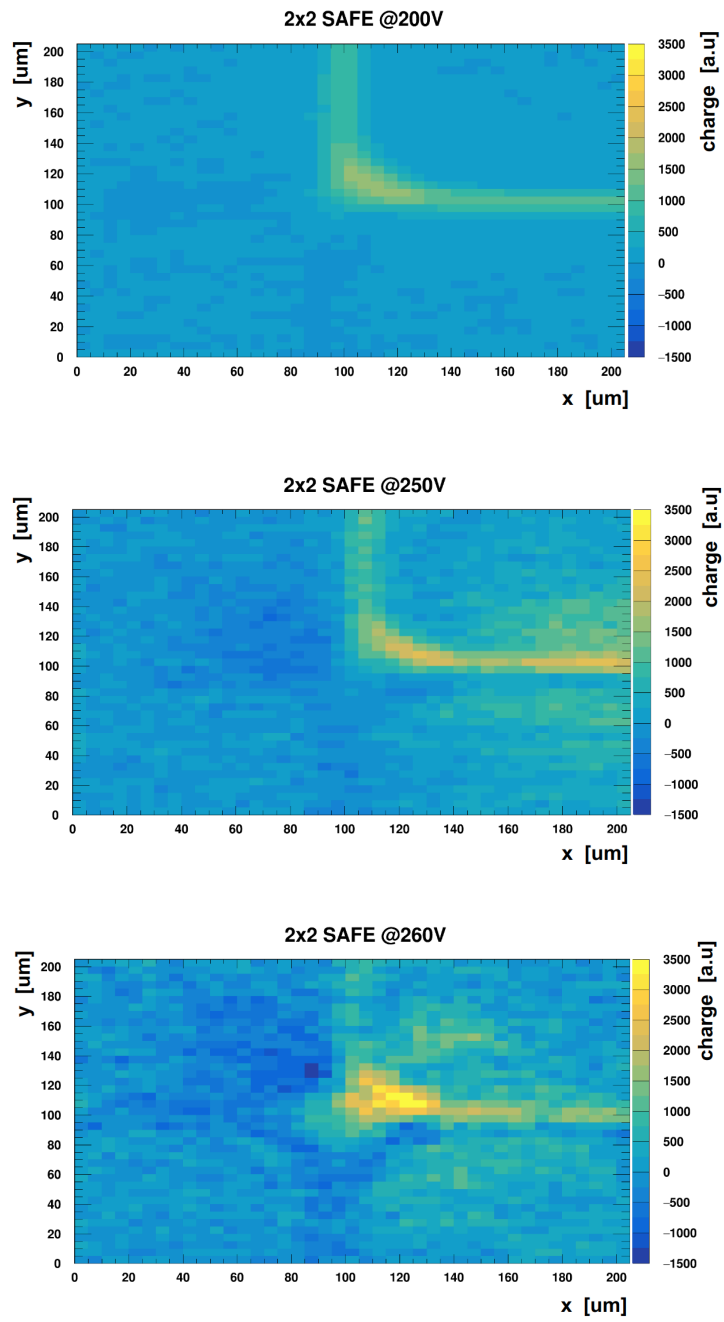


Figure 8: TCT 2D-map of the charge collected by Pad 1 at three different bias voltages: 200 V, 250 V, and 260 V.

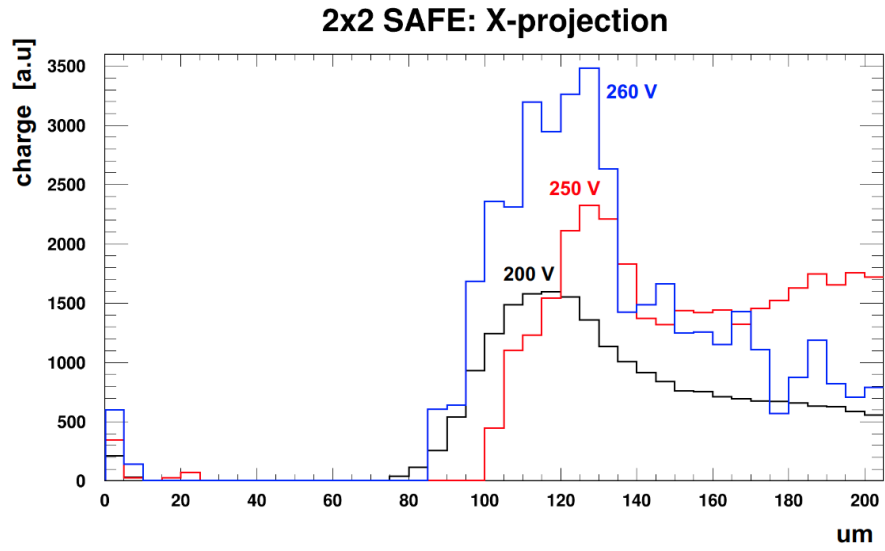


Figure 9: X-projections of the 2D-maps shown in figure 8 at  $y=100 \mu\text{m}$ . The current increase with bias at  $x = 120 \mu\text{m}$  signals the onset of the breakdown.

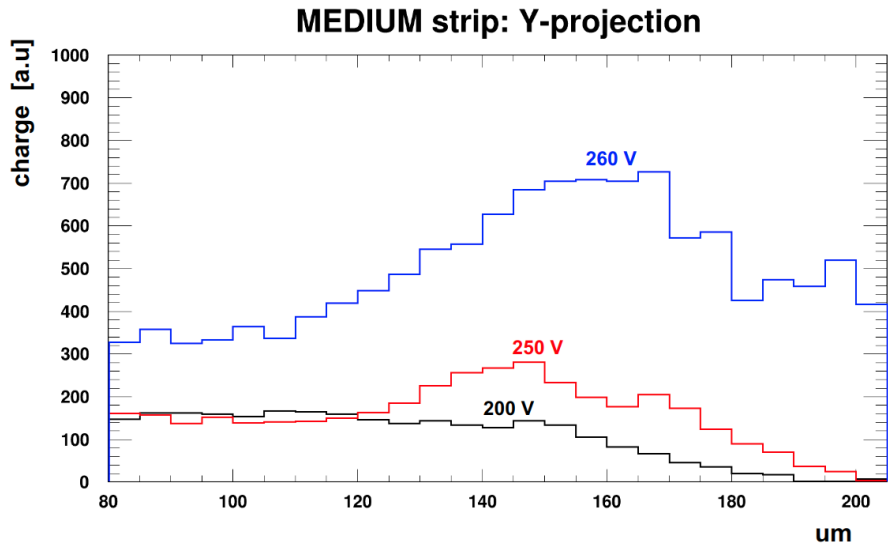


Figure 10: Charge collected in the inter-pad region by a strip sensor with Medium design for three different voltages.

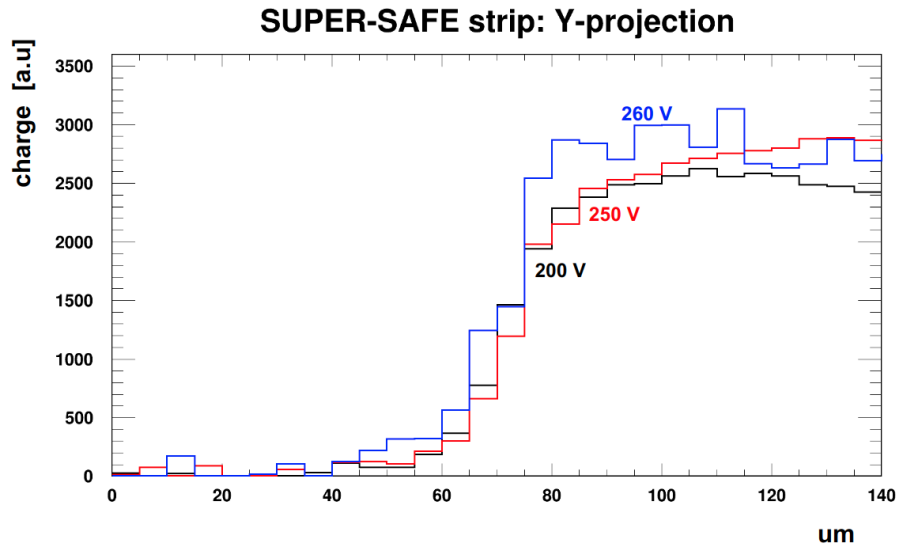


Figure 11: Charge collected in the inter-pad region by a strip sensor with Super-safe design for three different voltages.

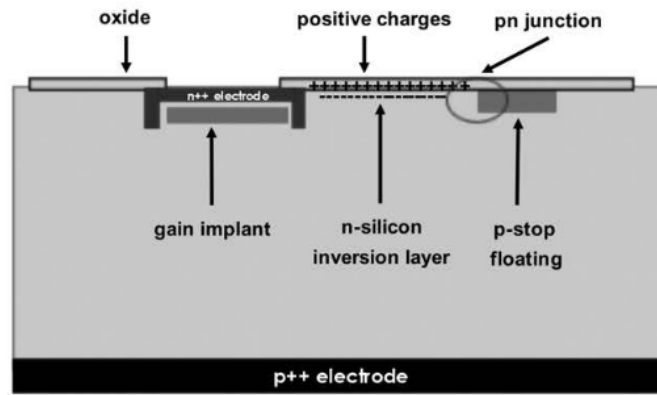


Figure 12: Schematic representation of the inversion layer establishing in the inter-pad region.

204 the oxide due to the positive charges present at the Si-SiO<sub>2</sub> interface. A sketch  
 205 is shown in Figure 12. The value of the *p*-stop doping determines how abrupt is  
 206 the *pn* junction and, consequently, how high the electric field is [18]. It follows  
 207 that high *p*-stop doping leads to higher fields, therefore, weakening the sensor  
 208 design.



209 The intensity of the electric field between the pixel and the  $p$ -stop is also  
 210 affected by their relative distance and the sensor thickness. The  $p$ -stop is elec-  
 211 trically floating, and it positions itself at a potential between the backplane  
 212 bias voltage and the ground level of the pixel. In thin sensors, the  $p$ -stop will  
 213 float to values closer to the backplane voltage, yielding a higher electric field in  
 214 the region pixel- $p$ -stop . In sensors with narrow inter-pad regions, this effect is  
 215 particularly important and can lead to an early breakdown.

216 To confirm the results obtained with the TCT analysis, complementary stud-  
 217 ies have been conducted using a CCD thermal camera. The area of the inter-pad  
 218 region scanned with the TCT has been captured with the ORCA2 camera at  
 219 different bias voltages. Figures 13 and 14 show pictures of a  $2 \times 2$  array with  
 220 *Safe* design, and of the strip device with *Super-safe* design, taken before and  
 221 after breakdown.

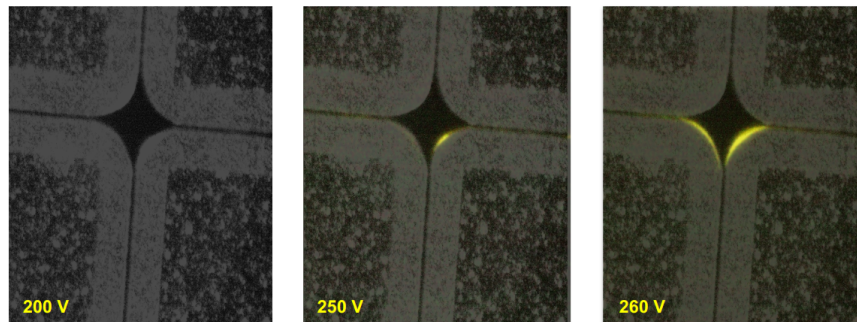


Figure 13: Pictures of the inter-pad region of the  $2 \times 2$  device with *Safe* design at three different voltages. The brighter area indicates the presence of high current density.

222 The *hot spots* (yellow regions) that can be seen in Figure 13 are regions  
 223 that emit visible photons due to the high current densities flowing through.  
 224 Such high densities are tied to the gain avalanche occurring while the sensor is  
 225 going into breakdown, as previously shown with the TCT scans. *Hot spots* are  
 226 particularly visible in correspondence to the corners, where the electric field is  
 227 higher. Similar pictures have been obtained from the devices with *Aggressive*  
 228 and *Medium* designs. As expected, the *Super-safe* device does not show any

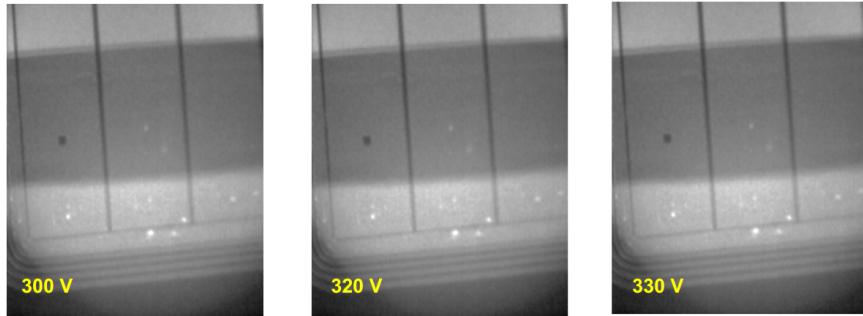


Figure 14: Pictures of the inter-pad region of the strip device with Super-safe design at three different voltages

229 signs of *hot spots*, see Figure 14.

#### 230 5.1. Micro-discharges in UFSD3 sensors

231 A second undesired effect has been observed on sensors of the UFSD3 pro-  
 232 duction: the appearance of large current spikes much before the breakdown  
 233 voltage, which prevents proper operation of the sensor as the dark count rate  
 234 increases significantly. Such large spikes have amplitude comparable to that of  
 235 real signals and are randomly distributed in time [16, 19, 20]. Figure 15 presents  
 236 a comparison between the normal baseline activity of a UFSD and that of a de-  
 237 vice affected by micro-discharges. The micro-discharges appear on both strip  
 238 and pad sensors; they do not depend on the sensor geometry or the inter-pad  
 239 designs and have been observed on both new and irradiated devices.

240 It is worth pointing out that the increase of the baseline activity naturally  
 241 happens in all UFSD devices a few volts before breakdown: it is an indica-  
 242 tion that the gain avalanche is going to start. What distinguishes the micro-  
 243 discharges in UFSD3 is they appear at a voltage much lower than  $V_{BD-Gain}$ .

244 To observe the micro-discharges, the DUT has been bonded on a custom  
 245 read-out board with all channels connected to an oscilloscope. The measure-  
 246 ments are performed inside a climate chamber, at +20 °C, with dry air fluxed.  
 247 The bias voltage is raised slowly in steps of 5-10 V until the breakdown occurs;  
 248 meanwhile, the baseline activity is monitored on the oscilloscope.

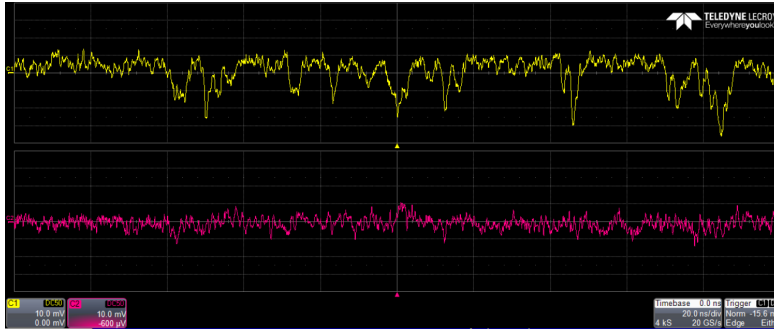


Figure 15: Comparison between a sensor with normal baseline activity (pink) and a sensor with micro-discharge (yellow). The vertical scale is 10 mV/division.

249 5.2. Conclusions of the measurement campaign on UFSD3

250 Two main issues have been identified in UFSD3:

- 251 • The sensors with *Aggressive*, *Medium* and *Safe* designs suffer from prema-  
252 ture breakdown.
- 253 • All designs show micro-discharges much before the breakdown, which pre-  
254 vents the proper operation of the sensors.
- 255 • The *Super-safe design* shows micro-discharges without going into prema-  
256 ture breakdown because the JTE and *p*-stop are relatively far away. De-  
257 spite that, its proper operation is compromised.

258 The TCT scans and the measurements with the ORCA2 camera demon-  
259 strated that the issues of the UFSD3 production are tied to the strong electric  
260 fields established in the inter-pad region between the JTE and the *p*-stop. Such  
261 electric fields are caused by:

- 262 • Highly doped *p*-stops
- 263 • A short distance between the JTE and the *p*-stops

## 264 6. Measurement campaign on UFSD3.1 sensors

265 The UFSD3.1 production was designed to explore further the properties  
266 of the UFSD inter-pad region. The measurement campaign started analyzing  
267 the  $I(V)$  characteristics of several devices: a summary is shown in Figures 16.  
268 Figure 16 (a) reports the  $I(V)$  characteristics of sensors with the same inter-  
269 pad design (Type 2) and different  $p$ -stop dopings, while figure 16 (b) reports  
270 the  $I(V)$  characteristics of sensors with the same  $p$ -stop doping and different  
271 inter-pad designs.

272 The breakdown voltage due to gain,  $V_{BD-Gain}$ , is expected to be 360-  
273 380 V for all sensors as they share the same gain layer dose: sensors going into  
274 breakdown earlier are considered to be suffering from premature breakdown.

275 Figure 16 (a) shows the strong influence of the  $p$ -stop doping on  $V_{BD}$ .  
276 Given that all sensors are of Type 2, the difference in  $V_{BD}$  is due solely to the  
277  $p$ -stop doping level. For high values of  $p$ -stop doping  $V_{BD}$  decreases consider-  
278 ably while for low values  $V_{BD}$  reaches its limiting value  $V_{BD-Gain}$ . Conversely,  
279 Figure 16 (b) shows the influence of the inter-pad design on  $V_{BD}$ , as all sen-  
280 sors have the same  $p$ -stop doping. The combination of the two plots clearly  
281 demonstrates the interplay of doping and geometry: a given type, for example  
282 Type 2, has an early breakdown if the  $p$ -stop is too doped, while a stronger  
283 design, for example Type 4, does not have an early breakdown even with a high  
284  $p$ -stop doping.

285 The ORCA2 camera detects *hot spots* much more rapidly than the TCT  
286 procedure, so it was used to test numerous devices. The measurements have  
287 been performed with the procedure described in Section 5.1. A few examples  
288 are shown in 17, 18. The remarkable feature of the ORCA2 testing campaign  
289 was that all sensors with premature breakdown show *hot spots* near the *Regions*  
290 *A/B* and not along the  $p$ -stop perimeter.

291 Figure 19, left side, shows that as the area of the region *Region A* increases,  
292  $V_{BD}$  decreases. Sensors with less-doped  $p$ -stops, those on W13, are less sensitive  
293 to this effect. A similar result is obtained when considering  $V_{BD}$  as a function

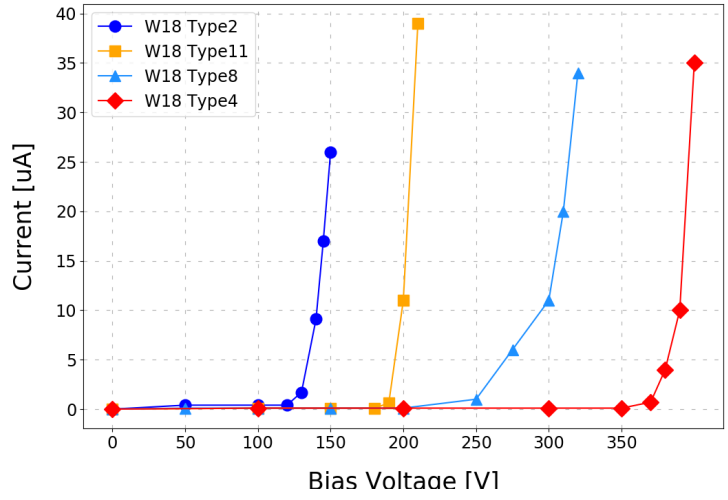
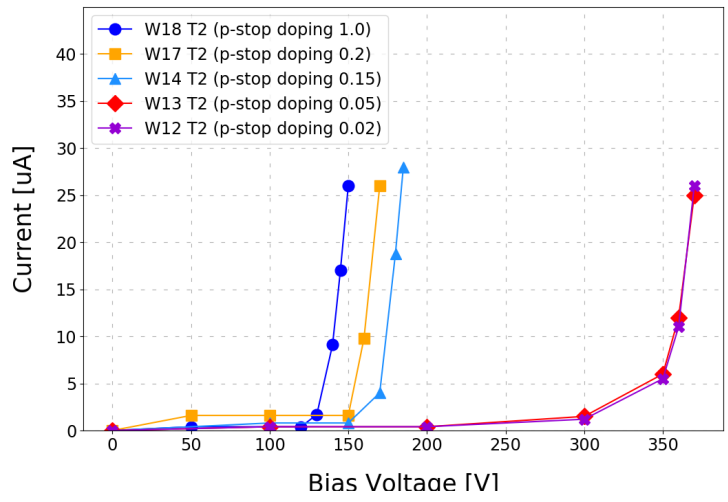


Figure 16: Top:  $I(V)$  of devices with different  $p$ -stop dopings having the same inter-pad design (Type 2). Bottom:  $I(V)$  of devices with different inter-pad design having the same  $p$ -stop doping.

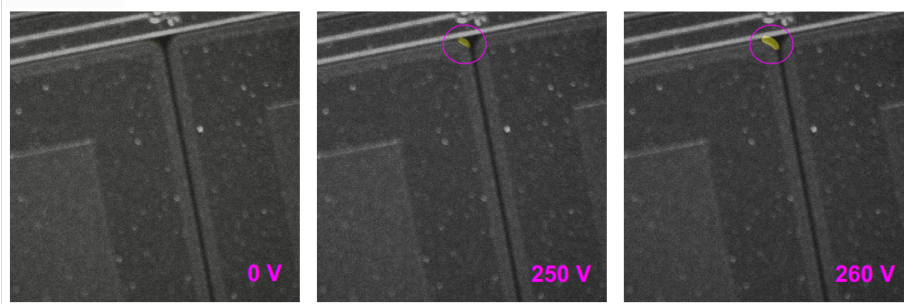


Figure 17: *Hot spots* in a Wafer18 Type 2 sensor near *Region B*

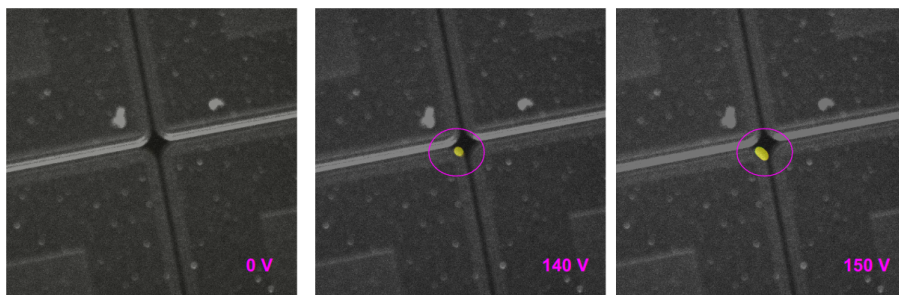


Figure 18: *Hot spots* in a Wafer14 Type 1 sensor near *Region A*

294 of the *Region B* area.

295 Figure 19, right side, instead, highlights that a less-doped  $p$ -stop allows to  
 296 reach a higher  $V_{BD}$ . Types 3 and 4 have a low dependence on the  $p$ -stop  
 297 doping because they have small *Region A*, *B* areas; Type 10, instead, is not  
 298 susceptible to the  $p$ -stop doping because of its grid guard-ring design. Type 1  
 299 has small  $p$ -structures but its  $V_{BD}$  depends strongly on the  $p$ -stop doping since  
 300 it features the most aggressive design. The figure also shows that sensors from  
 301 the wafers with lowest  $p$ -stop dopings (W12 and W13) have a  $V_{BD}$  that reaches  
 302 the limiting value of 360-380 V.

303 In wafers 12 and 13 the  $p$ -stop dose is such that the breakdown always  
 304 happens due to internal gain, regardless of the inter-pad design. Wafer 14  
 305 shows a similar trend, with only the most aggressive designs going into an early

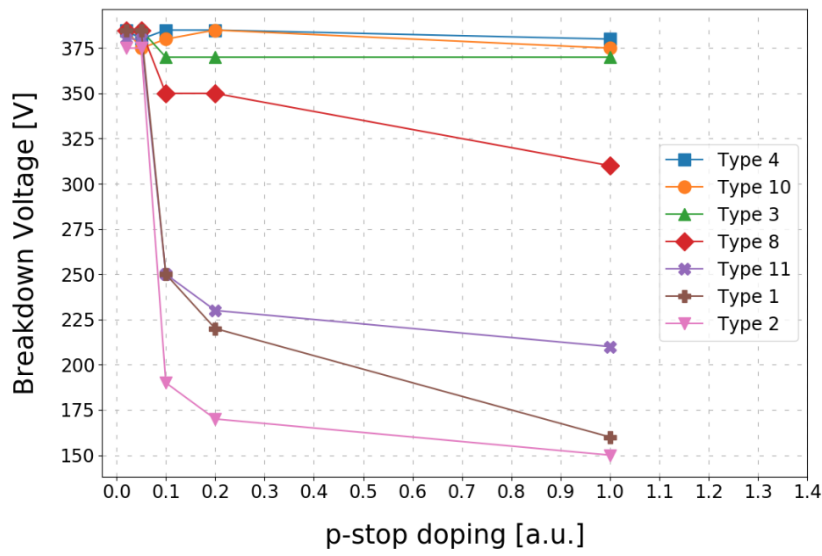
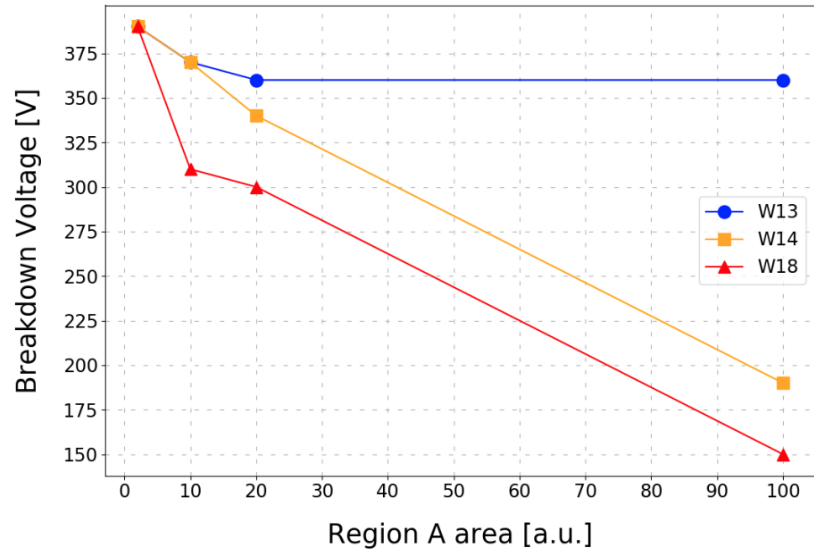


Figure 19: Top:  $V_{BD}$  as a function of *Region A* area for different Wafers (different *p*-stop doping). Bottom:  $V_{BD}$  as a function of *p*-stop doping for different sensor types.

306 breakdown.

307 The results of this measurement campaign can be summarized as follows:

- 308 • Premature breakdown occurs in the inter-pad region near the pad corners.
- 309 • Premature breakdown depends both on the  $p$ -stop doping and on the  
310 inter-pad design/width.
- 311 • Corners (*Regions A, B*) are the most critical areas.
- 312 • Large and/or too-doped  $p$ -structures lead to premature breakdown, pre-  
313 venting proper operation of sensors.
- 314 • The combination of low-doped  $p$ -stop and small-area  $A, B$  regions is the  
315 optimal choice to avoid early breakdown.
- 316 • An inter-pad width of  $\sim 25 \mu\text{m}$  is achievable without incurring in  
317 premature breakdown (UFSD3.1 Type 4), adopting low-doped  $p$ -stops.  
318 Sensors with narrower regions (Type 1) are unstable.

#### 319 *6.1. The effect of floating pads on $V_{BD}$ .*

320 The breakdown voltage of several UFSD3.1  $2 \times 2$  devices with 0, 1, and 2  
321 floating pads has been measured for several wafers. Figure 20 left side, shows  
322 that sensors from W13 (low  $p$ -stop doping) are almost insensitive to floating  
323 pads, whereas, Figure 20 right side, most sensors from W18 (high  $p$ -stop dop-  
324 ing) have a breakdown voltage decreasing with the number of floating pads.  
325 This study confirms the key importance of a low-doped  $p$ -stop to assure stable  
326  $V_{BD}$  even in non-standard working conditions such as those with one or more  
327 floating pads. Type 10, due to its grid guard-ring design, is very resilient to  
328 floating pads, regardless of the  $p$ -stop doping.

#### 329 *6.2. Micro-discharges in UFSD3.1*

330 UFSD3.1 Wafers 13 and 14 have been tested for the micro-discharges ef-  
331 fect, following the same procedure described for the UFSD3 production, see  
332 Section 5.1. The results are reported in Tables 5 and 6.



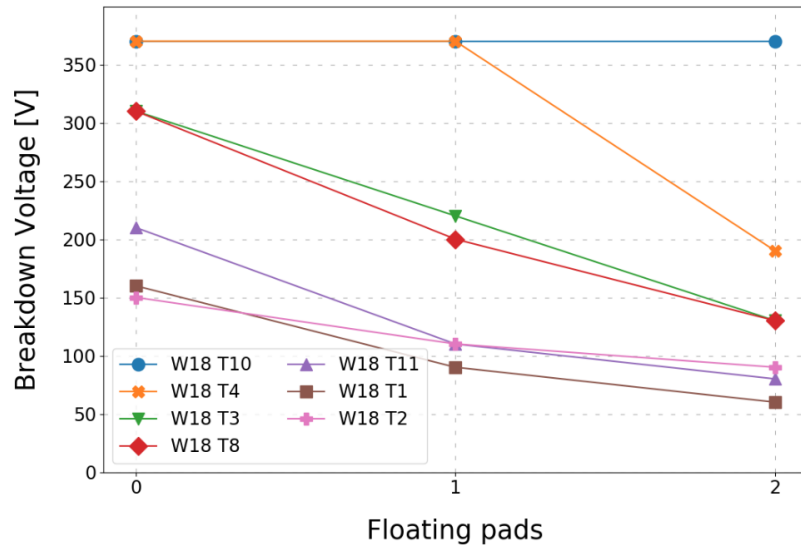
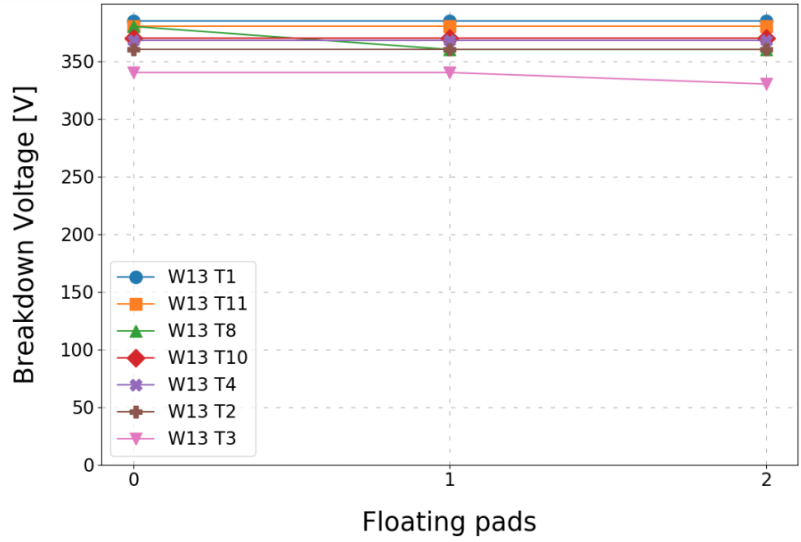


Figure 20: Top:  $V_{BD}$  as a function of the number of floating pads for different sensor designs from Wafer 13. Types 4 and 10 have the same  $V_{BD}$  (370 V), but the curves have been slightly shifted for illustration purposes. Bottom:  $V_{BD}$  as a function of the number of floating pads for different sensor designs from Wafer 18.

	0 pad floating			1 pad floating		
Type	$V_{BD}$	Discharge	$V_{Discharge}$	$V_{BD}$	Discharge	$V_{Discharge}$
Type 1	385	NO		350	NO	
Type 8	380	NO		380	NO	
Type 9	365	NO		355	NO	
Type 10	370	NO		335	NO	
Type 11	380	NO		335	NO	

Table 5: Results on the micro-discharges effect on UFSD3.1 W13.

	0 pad floating			1 pad floating		
Type	$V_{BD}$	Discharge	$V_{Discharge}$	$V_{BD}$	Discharge	$V_{Discharge}$
Type 1	260	YES	245	170	YES	145
Type 8	350	NO		350	NO	
Type 9	370	NO		310	NO	
Type 10	380	NO		380	NO	
Type 11	250	NO		185	NO	

Table 6: Results on the micro-discharges effect on UFSD3.1 W14.

333 Wafer 13 does not show any signs of discharges, even with one pad floating;  
334 whereas wafer 14 has some, occurring 15-25 V before the breakdown, but only in  
335 the Type 1 design, which is the most aggressive. Micro-discharges are, therefore,  
336 not an issue for UFSD3.1 wafers with low  $p$ -stop doping since they are present  
337 only in the most aggressive design of W14, and only very close to breakdown,  
338 not affecting the device operation.

339 Radiation damage has two main effects that can influence the presence of  
340 micro-discharges: (i) the sensor is operated at higher bias voltage; (ii) the *ac-*  
341 *ceptor removal* mechanism [21] decreases the doping of the  $p$ -stop structures.  
342 The two effects act in opposite directions: (i) leads to an increase in the electric  
343 field, whereas (ii) lowers it. Table 7 reports the results obtained on wafer 14,  
344 irradiated at  $\phi = 4 \cdot 10^{14}$  n<sub>eq</sub>/cm<sup>2</sup> and  $\phi = 8 \cdot 10^{14}$  n<sub>eq</sub>/cm<sup>2</sup>. The sensors have

345 been irradiated, without bias, with neutrons at the JSI TRIGA research reactor  
 346 in Ljubljana [22]. Types 1 and 11 show micro-discharges at  $\phi = 4 \cdot 10^{14}$  n<sub>eq</sub>/cm<sup>2</sup>,  
 347 whereas only Type 1 features this effect at  $\phi = 8 \cdot 10^{14}$  n<sub>eq</sub>/cm<sup>2</sup>. Similar results  
 348 are obtained in the "1 floating pad" configuration: only types 1 and 11 show  
 349 signs of micro-discharges. Interestingly, at  $\phi = 1.5 \cdot 10^{15}$  n<sub>eq</sub>/cm<sup>2</sup>, all types  
 350 break down above 600 V, and none has micro-discharges before  $V_{BD}$  .

	$\phi = 4 \cdot 10^{14}$ n <sub>eq</sub> /cm <sup>2</sup>			$\phi = 8 \cdot 10^{14}$ n <sub>eq</sub> /cm <sup>2</sup>		
Type	$V_{BD}$	Discharge	$V_{Discharge}$	$V_{BD}$	Discharge	$V_{Discharge}$
Type 1	390	YES	270	440	YES	390
Type 8	470	NO		510	NO	
Type 9	470	NO		540	NO	
Type 10	500	NO		590	NO	
Type 11	430	YES	300	530	NO	

Table 7: Results on the micro-discharges effect on UFS3.1 W14 irradiated.

351 Radiation damage, therefore, does not trigger the presence of micro-discharges,  
 352 even if the sensors are operated at a much higher bias voltage.

### 353 6.3. W14 inter-pad resistance

354 The main task of the *p*-stop implant is to assure pixel isolation by interrupt-  
 355 ing the charge inversion layer, as shown in Figure 12. This can only happen if  
 356 the *p*-stop doping level is high enough. It is, therefore, mandatory to check that  
 357 inter-pad resistance remains high at all *p*-stop dose levels considered, before and  
 358 after irradiation.

359 The chosen figure of merit that quantifies the pad isolation is the inter-pad  
 360 resistance, namely the resistance of a pad to ground when all the other pads  
 361 and the guard-ring are connected to ground. The measurements have been  
 362 performed on 2×2 pad arrays of types 8, 9, and 10, at +20 °C. Three pads and  
 363 the guard-ring are grounded, while a voltage sweep between -10 V and +10 V is  
 364 performed on the pad under test, as shown in Figure 21. The measured current

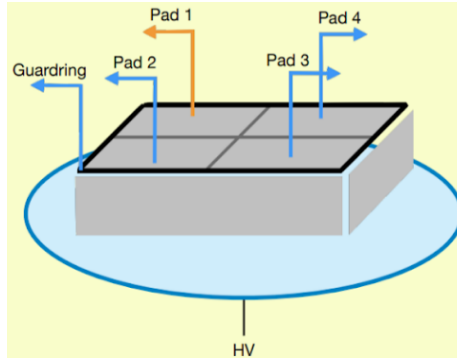


Figure 21: Sketch of the setup used to measure the inter-pad resistance.

365 is plotted as a function of the bias voltage applied to the pad: the slope of the  
 366 curve provides the inter-pad resistance. That is actually a lower limit, since the  
 367 measurement does not account for the bulk current, which should be subtracted  
 368 from the measured current, thus yielding a larger inter-pad resistance.

369 The measurement is performed on a pre-irradiation sensor and repeated  
 370 on devices irradiated at  $\phi = 4 \cdot 10^{14} \text{ n}_{\text{eq}}/\text{cm}^2$ ,  $\phi = 8 \cdot 10^{14} \text{ n}_{\text{eq}}/\text{cm}^2$ ,  $\phi =$   
 371  $1.5 \cdot 10^{15} \text{ n}_{\text{eq}}/\text{cm}^2$ , and  $\phi = 3 \cdot 10^{15} \text{ n}_{\text{eq}}/\text{cm}^2$ .

372 The results are shown in Figure 22: the resistance lowers with increasing  
 373 exposure, but it remains high, about  $10 \text{ G}\Omega$ , even at the highest fluence, proving  
 374 that the pad is well isolated. As a comparison, the UFSDs that will instru-  
 375 ment the CMS Endcap Timing layer must have an inter-pad resistance higher  
 376 than  $0.1 \text{ G}\Omega$  up to a radiation fluence of  $1.5 \cdot 10^{15} \text{ n}_{\text{eq}}/\text{cm}^2$  [23].

## 377 7. Conclusions

378 The development of LGADs array with high fill factor requires the study  
 379 of the property of the inter-pad region. Two FBK productions, UFSD3 and  
 380 UFSD3.1 have been dedicated to this analysis.

381 The main conclusion of the studies presented in the paper is that a high  
 382  $p$ -stop doping leads to early breakdown and micro-discharges. When using a

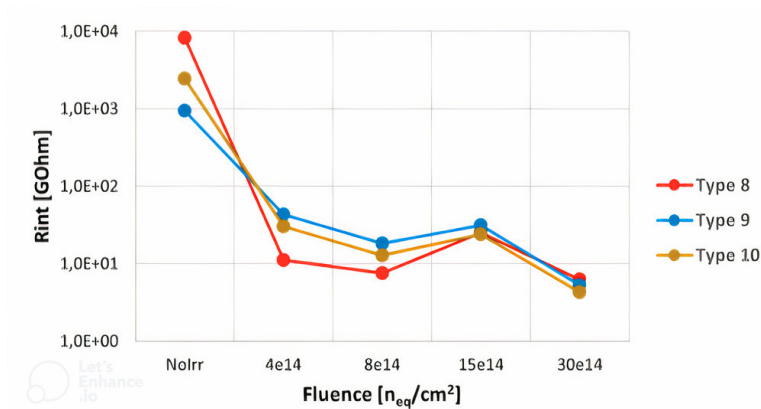


Figure 22: Inter-pad resistance of UFSD3.1 W14 as a function of fluence.

383 *p*-stop grid and low *p*-stop doping, inter-pad distances as small as 25  $\mu\text{m}$  are  
 384 achievable.

385 UFSDs with a low-doped *p*-stop, and small-area structures in the inter-  
 386 pad region have also been proven to be rather insensitive to floating pads and  
 387 irradiation up to a fluence of  $1.5 \cdot 10^{15} \text{ n}_{eq}/\text{cm}^2$ . Pad isolation is not an issue  
 388 either, even for highly irradiated devices.

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